NETWORK ANALYSIS
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INTRODUCTION

Analysis and Design:—continuous processes for Improvement of response ---- (Basis of Research and Development activities)

Analysis

Input

or Excitation
(known)

or Response
(given)

or Output
(to be obtained)

Pre requisites: (i) A.C single phase circuits - chapter 2-ELE 15/25
(ii) Elementary Calculus-Part B & C - Mat 11
(iii) Differential Equations part C - MAT 21
(iv) Laplace Transform -part D – MAT 21
(v) Solutions of Simultaneous equations by Kramars Rule
(vi) Simple Matrix operations with real numbers

Co-requisite: usage of calculator (preferably CASIO fx 570ms or fx 991ms)

Books for Reference:
(i) Engineering circuit Analysis----- Hayt, Kimmerily and Durbin for chapters 1,3,4,6,7
(ii) Network Analysis--- Van ValkenBerg- chapters 5,6,7
(iii) Network and Systems---- Roy Choudary - chapter 2
**V-I RELATIONS**

<table>
<thead>
<tr>
<th>ELEMENT</th>
<th>Voltage</th>
<th>Current</th>
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</thead>
<tbody>
<tr>
<td></td>
<td>t-Domain</td>
<td>Jw-Domain</td>
</tr>
<tr>
<td>RESISTANCE (R)</td>
<td>$v = Ri$</td>
<td>$V = RI$</td>
</tr>
<tr>
<td>INDUCTANCE (L)</td>
<td>$v = L(di/dt)$</td>
<td>$V = (JwL)i$</td>
</tr>
<tr>
<td>CAPACITANCE (C)</td>
<td>$V = (1/C)\int (idt)$</td>
<td>$V = (-J/wC)i$</td>
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</table>
| $X_L = wL$       | $X_C = 1/wC$ | For $t 

**BASIC LAWS**

1. **OHMS LAW**
   
   $$V = IZ$$
   
   $I_{AB}$-Current from A to B
   
   $V_{AB}$=Voltage of A w.r.t B

2. **KCL**
   
   $$i_1 + i_4 + i_5 = i_2 + i_3$$
   
   $$\sum i = 0$$ algebraic sum
   
   or $$\sum i_{in} = \sum i_{out}$$ ($I_{in} = -I_{out}$)

3. **KVL**
   
   $$\sum v = 0$$ algebraic sum
   
   $$\sum v_{rise} = \sum v_{drop}$$ ($V_{rise} = -V_{drop}$)
   
   $E_1 - E_2 = V_1 - V_2 + V_3 - V_4 = I_1Z_1 - I_2Z_2 + I_3Z_3 - I_4Z_4$
## CONNECTIONS

<table>
<thead>
<tr>
<th>SERIES</th>
<th>PARELLEL</th>
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<tbody>
<tr>
<td><img src="image" alt="Series Connections Diagram" /></td>
<td><img src="image" alt="Parallel Connections Diagram" /></td>
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Voltage Division

\[ V_i = \left(\frac{Z_i}{Z}\right)V \]

\[ I = \frac{V}{Z} = \frac{V_1}{Z_1} = \frac{V_2}{Z_2} = \cdots \]

Current Division

\[ Y = \sum_{i=1}^{n} Y_i = Y_1 + Y_2 + Y_3 + \cdots + Y_n \]

\[ I_1 = \left(\frac{Y}{Y_i}\right)I \]

\[ V = \frac{I}{Y} = \frac{I_1}{Y_1} = \frac{I_2}{Y_2} = \cdots \]
### Problems

1. Calculate the voltages $V_{12}, V_{23}, V_{34}$ in the network shown in Fig, if $V_a=17.32+j10\ V$, $V_b=30\ 80^\circ$ and $V_C=15\ -100^\circ$ with Calculator in complex and degree mode.

   $V_{12} = -V_c + V_b = (0-15\ -100\ +30\ 80^\circ) = 45\ 80^\circ \ V$ *

   $V_{23} = V_a-V_b+V_c = V_a - V_{12} - V_c = 17.32+10i - 45\ 80^\circ = 35.61-74.52^\circ$

   $V_{34} = V_b - V_a = 30\ 80^\circ - 17.32-10i = 23\ 121.78^\circ$

2. How is current of 10A shared by 3 impedances $Z_1=2-J5\ \Omega$, $Z_2 = 6.708\ 26.56$ and $Z_3 = 3 + J4$ all connected in parallel.

   Ans:

   $Z = Y^{-1} = ((2-5i)^{-1} + (6.708\ 26.56)^{-1} + (3+4i)^{-1} = 3.06\ 9.55^\circ$

   $V=IZ = 30.6\ 9.55^\circ \ I_1 = V/Z_1=(30.6\ 9.55^\circ ) - (2-5i) = 5.68 - 77.75^\circ$

   $I_2 = V/Z_2 = (30.6\ 9.55^\circ ) - (6.708\ 26.56) = 4.56 - 17^\circ$

   $I_3 = V/Z_2 = (30.6\ 9.55^\circ ) - (3+4i) = 6.12\ -43.6^\circ$

3. In the circuit determine what voltage must be applied across AB in order that a current of 10 A may flow in the capacitor.

   $V_{AC} = (7-8i)(10) = 106.3\ -48.8^\circ$

   $I_1 = V_{AC} = 13.61\ -99^\circ$

   $I = I_1+I_2 = 10\ 0^\circ + 13.6\ -99^\circ = 15.57\ -59.66^\circ$

   $V = V_1+V_2 = 106.3\ -48.8 + (15.576\ -59.66) (8+10i)=289\ -22^\circ$
Network analysis

Network is a system with interconnected electrical elements. Network and circuit are the same. The only difference being a circuit shall contain at least one closed path.

Network Elements

**Sources**
- **Independent Sources**
  - Voltage Source (ideal)
  - Current Source (ideal)

**Dependant Sources**
- (a) Current controlled current source
- (b) Voltage controlled current source
- (c) Voltage controlled voltage source
- (d) Current controlled voltage source

**Passive Elements**
- **R** (Energy Consuming Element)
- **L** (Energy storing element in a magnetic field)
- **C** (Energy storing element in an Electric field)

**Electrical Elements**

(Value of source Quantity is not affected in anyway by activities in the reminder of the circuit.)

(Source quantity is determined by a voltage or current existing at some other Location in the circuit) These appear in the equivalent models for many electronic devices like transistors, OPAMPS and integrated circuits.)
**TERMINOLOGY**

- **Voltage controlled current source**
- **Node (Junction)**
- **Current controlled Voltage source**
- **Reference node**
- **Practical current source**
- **Mesh (loop)**
- **Loop**
- **Practice Voltage source**
TYPES OF NETWORKS

Linear and Nonlinear Networks:
A network is linear if the principle of superposition holds i.e if e1(t), r1 (t) and e2(t), r2 (t) are excitation and response pairs then if excitation is e1 (t) + e2 (t) then the response is r1 (t) + r2(t).

The network not satisfying this condition is nonlinear
Ex:- Linear – Resistors, Inductors, Capacitors.
Nonlinear – Semiconductors devices like transistors, saturated iron core inductor, capacitance of a p-n function.

Passive and active Networks:
A Linear network is passive if (i) the energy delivered to the network is nonnegative for any excitation. (ii) no voltages and currents appear between any two terminals before any excitation is applied.
Example:- R,L and C.

Active network:- Networks containing devices having internal energy – Generators, amplifiers and oscillators.

Unilateral & Bilateral:
The circuit, in which voltage current relationship remains unaltered with the reversal of polarities of the source, is said to be bilateral.
Ex:- R, L & C

If V-I relationships are different with the reversal of polarities of the source, the circuit is said to be unilateral.
Ex:- semiconductor diodes.

Lumped & Distributed:
Elements of a circuit, which are separated physically, are known as lumped elements.
Ex:- L & C.

Elements, which are not separable for analytical purposes, are known as distributed elements.
Ex:- transmission lines having R, L, C all along their length.

In the former case Kirchhoff’s laws hold good but in the latter case Maxwell’s laws are required for rigorous solution.

Reciprocal:
A network is said to be reciprocal if when the locations of excitation and response are interchanged, the relationship between them remains the same.
Source Transformation

In network analysis it may be required to transform a practical voltage source into its equivalent practical current source and vice versa. These are done as explained below.

Consider a voltage source and a current source as shown in Figure 1 and 2. For the same load $Z_L$ across the terminals a & b in both the circuits, the currents are

$$I_L = \frac{E_S}{Z_S + Z_L} \quad \text{in fig 1} \quad \text{and} \quad I_L = \frac{I_S Z_P}{Z_P + Z_L} \quad \text{in fig 2}$$

For equivalence

$$E_S = I_S \frac{Z_P}{Z_P + Z_L}$$

Therefore $E_S = I_S \frac{Z_P}{Z_P + Z_L}$ and $Z_S = Z_P$

Therefore

$$I_S = \frac{E_S}{Z_P} = \frac{E_S}{Z_S}$$

Transformation from a practical voltage source to a practical current source eliminates a node. Transformation from a practical current source to a current source eliminates a mesh.

A practical current source is in parallel with an impedance $Z_p$ is equivalent to a voltage source $E_s = I_s Z_p$ in series with $Z_p$.

A practical voltage source $E_s$ in series with a impedance $Z_s$ is equivalent to a current source $E_s/Z_s$ in parallel with $Z_s$. 
**SOURCE SHIFTING**

Source shifting is occasionally used to simplify a network. This situation arises because of the fact that an ideal voltage source cannot be replaced by a current source. Likewise, an ideal current source cannot be replaced by a voltage source. But such a source transformation is still possible if the following techniques are followed.

(a) E shift operation

(b) I shift operation
Sources with equivalent terminal characteristics

(i) Series voltage sources

(ii) Parallel voltage sources (ideal)

(iii) Parallel current sources

(iv) Series current sources (ideal)

(v) Voltage source with parallel Z

(vi) Current source with series Z

(vii) V and I in Parallel

(viii) V and I in Series

1. Any element in parallel with ideal voltage source (dependent or independent) is trivial
2. Any element in series with ideal current source (dependent or independent) is trivial
Clues to Simplify the Network
(A network with too many trivial elements)

Fig:
Network of fig with trivial elements marked by \( \times \)
Network after removal of trivial elements
**Delta-star transformation**

A set of star connected (Y or T) immittances can be replaced by an equivalent set of mesh (Δ or π) connected immittances or vice versa. Such a transformation is often necessary to simplify passive networks, thus avoiding the need for any mesh or nodal analysis.

For equivalence, the immittance measured between any two terminals under specified conditions must be the same in either case.

**Δ to Y transformation:**

Consider three Δ-connected impedances \( Z_{AB}, Z_{BC} \) and \( Z_{CA} \) across terminals A, B and C. It is required to replace these by an equivalent set \( Z_A, Z_B \) and \( Z_C \) connected in star.

![Diagram](https://via.placeholder.com/150)

In Δ, impedance measured between A and B with C open is

\[
\frac{Z_{AB}(Z_{BC} + Z_{CA})}{Z_{AB} + Z_{BC} + Z_{CA}}
\]

With C open, in Y, impedance measured between A and B is \( Z_A + Z_B \).

For equivalence \( Z_A + Z_B = \frac{Z_{AB}(Z_{BC} + Z_{CA})}{Z_{AB} + Z_{BC} + Z_{CA}} \)  

Similarly for impedance measured between B and C with A open

\[
Z_B + Z_C = \frac{Z_{BC}(Z_{CA} + Z_{AB})}{Z_{AB} + Z_{BC} + Z_{CA}}
\]

For impedance measured between C and A with B open

\[
Z_C + Z_A = \frac{Z_{CA}(Z_{AB} + Z_{BC})}{Z_{AB} + Z_{BC} + Z_{CA}}
\]

Adding (1), (2) and (3)

\[
2 (Z_A + Z_B + Z_C) = \frac{2 (Z_{AB}Z_{BC} + Z_{BC}Z_{CA} + Z_{CA}Z_{AB})}{Z_{AB} + Z_{BC} + Z_{CA}}
\]

\[
Z_A = \frac{(Z_{AB}Z_{BC} + Z_{BC}Z_{CA} + Z_{CA}Z_{AB})}{Z_{AB} + Z_{BC} + Z_{CA}} - (Z_B + Z_C)
\]

Substituting for \( Z_B + Z_C \) from (2)

\[
Z_A = \frac{Z_{CA}Z_{AB}}{Z_{AB} + Z_B + Z_{CA}} = \frac{Z_{CA} Z_{AB}}{\sum Z_{AB}}
\]

\[
Z_A = \frac{Z_{AB} Z_{BC}}{\sum Z_{AB}}
\]
Similarly by symmetry  \( Z_B = \ldots \)

\[
Z_C = \frac{Z_{BC} Z_{CA}}{\sum Z_{AB}}
\]

If \( Z_{AB} = Z_{BC} = Z_{CA} = Z_\Delta \) then \( Z_A = Z_B = Z_C = Z_Y = \frac{Z_\Delta}{3} \).

**Y to Δ transformation:**

Consider three Y connected admittance \( Y_a, Y_b \) and \( Y_c \) across the terminals A, B and C. It is required to replace them by a set of equivalent Δ admittances \( Y_{ab}, Y_{bc} \) and \( Y_{ca} \).

Admittance measured between A and B with B & C shorted

\[
\frac{Y_A(Y_B + Y_C)}{Y_A+Y_B+Y_C}
\]

In Δ \( Y_{AB} + Y_{CA} \)

For equivalence \( Y_{AB} + Y_{CA} = \frac{Y_A(Y_B + Y_C)}{Y_A+Y_B+Y_C} \) \( \ldots \) \( (1) \)

Admittance between B and C with C & A shorted

\[
\frac{Y_B(Y_C + Y_A)}{Y_A+Y_B+Y_C}
\]

Admittance between C and A with A & B shorted

\[
\frac{Y_C(Y_A + Y_B)}{Y_A+Y_B+Y_C}
\]

Adding (1), (2) and (3) \( Y_{AB} + Y_{BC} + Y_{CA} \)

\[
\frac{Y_A Y_B + Y_B Y_C + Y_C Y_A}{Y_A+Y_B+Y_C}
\]

\[
Y_{AB} = \frac{\sum Y_A Y_B}{\sum Y_A} - (Y_{BC} + Y_{CA})
\]

substituting from (3)

\[
Y_{BC} = \frac{Y_B Y_C}{Y_A+Y_B+Y_C} : Y_{CA} = \frac{Y_A Y_B}{Y_A+Y_B+Y_C}
\]

In terms of impedances,

\[
Z_{AB} = \frac{Y_A + Y_B + Y_C}{Y_A Y_B} = \frac{Z_A Z_B + Z_B Z_C + Z_C Z_A}{Z_C}
\]

Similarly \( Z_{BC} = \frac{Z_A Z_B + Z_B Z_C + Z_C Z_A}{Z_B} \)

\[
Z_A Z_B + Z_B Z_C + Z_C Z_A
\]

\[
Z_A Z_B + Z_B Z_C + Z_C Z_A
\]
\[ Z_{CA} = \ldots \]

If \( Z_A = Z_B = Z_C = Z_Y \) then \( Z_{AB} = Z_{BC} = Z_{CA} = Z = 3Z_Y \).

**NETWORK THEOREMS**

Mesh current or node voltage methods are general methods which are applicable to any network. A number of simultaneous equations are to be set up. Solving these equations, the response in all the branches of the network may be attained. But in many cases, we require the response in one branch or in a small part of the network. In such cases, we can use network theorems, which are the aides to simplify the analysis. To reduce the amount of work involved by considerable amount, as compared to mesh or nodal analysis. Let us discuss some of them.

**SUPERPOSITION THEOREM**

The response of a linear network with a number of excitations applied simultaneously is equal to the sum of the responses of the network when each excitation is applied individually replacing all other excitations by their internal impedances.

Here the excitation means an independent source. Initial voltage across a capacitor and the initial current in an inductor are also treated as independent sources.

This theorem is applicable only to linear responses and therefore power is not subject to superposition.

During replacing of sources, dependent sources are not to be replaced. Replacing an ideal voltage source is by short circuit and replacing an ideal current source is by open circuit.

“In any linear network containing a number of sources, the response (current in or voltage across an element) may be calculated by superposing all the individual responses caused by each independent source acting alone, with all other independent voltage sources replaced by short circuits and all other independent current sources replaced by open circuits”. Initial capacitor voltages and initial inductor currents, if any, are to be treated as independent sources.

To prove this theorem consider the network shown in fig.

We consider only one-voltage sources and only one current sources for simplicity. It is required to calculate \( I_a \) with \( I_s \) acting alone the circuit becomes

\[
\begin{align*}
\frac{I_s}{Z_1 + Z_2 + Z_3 Z_4} & \quad \frac{Z_3}{Z_1 + Z_2 + Z_3 Z_4} \\
\frac{Z_3}{Z_1 + Z_2 + Z_3 Z_4} & \quad \frac{Z_4}{Z_1 + Z_2 + Z_3 Z_4}
\end{align*}
\]
\[ I_{a1} = \frac{Z_1 Z_3}{(Z_1 + Z_2 + Z_3) Z_4 + (Z_1 + Z_2) Z_3} \]  
\[ \text{with } E_S \text{ acting alone} \]  
\[ I_{a1} = \frac{-E_S}{Z_4 + (Z_1 + Z_2 + Z_3) Z_3} \]  
\[ \text{Next converting the current source to voltage source, the loop equations} \]  
\[ I_2 = \frac{Z_1 + Z_2 + Z_3}{Z_1 + Z_2 + Z_3} - \frac{Z_3}{Z_3 + Z_4} \]  
\[ \text{From equation (1), (2) and (3) } I_{a1} + I_{a2} = I_2 = I_a \]  
\[ \text{Hence proof} \]  

**Reciprocity Theorem:**
In an initially relaxed linear network containing one independent source only. The ratio of the response to the excitation is invariant to an interchange of the position of the excitation and the response.

i.e if a single voltage source Ex in branch X produces a current response Iy the branch Y, then the removal of the voltage source from branch x and its insertion in branch Y will produce the current response Iy in branch X.

Similarly if the single current source Ix between nodes X and X' produces the voltage response Vy between nodes Y and Y' then the removal of the current source from X and X' and its insertion between Y and Y' will produce the voltage response Vy between the nodes X and X'.

Between the excitation and the response, one is voltage and other is current. It should be noted that after the source and response are interchanged, the current and the voltages in other parts of the network will not remain the same.

**Proof:**

Consider a network as shown in which the excitation is E and the response is I in Z4. The reading of the ammeter is

\[
I_1 = \frac{E}{Z_1 + Z_3 (Z_2 + Z_4)} \cdot \frac{Z_3}{Z_2 + Z_3 + Z_4}
\]
\[ I_1 = \frac{E \cdot Z_3}{Z_1 \left( Z_2 + Z_3 + Z_4 \right) + Z_3 \left( Z_2 + Z_4 \right)} \quad \cdots (1) \]

Next interchange the source and ammeter.

Now the reading of the Ammeter is:

\[ I_2 = \frac{E}{(Z_2 + Z_4) + Z_1 \cdot Z_3} \cdot \frac{Z_3}{Z_1 + Z_3} \]

\[ I_2 = \frac{E \cdot Z_3}{Z_1 \left( Z_2 + Z_3 + Z_4 \right) + Z_3 \left( Z_2 + Z_4 \right)} \quad \cdots (2) \]

From (1) & (2)

\[ I_1 = I_2 \]

It can be similarly be shown for a network with current sources by writing node equations.
**Transfer Impedance:**

The transfer impedance between any two pairs of terminals of a linear passive network is the ratio of the voltage applied at one pair of terminals to the resulting current at the other pair of terminals.

With this definition the reciprocity theorem can be stated as:

“Only one value of transfer impedance is associated with two pairs of terminals of a linear passive network”.

![Diagram showing transfer impedance between terminals](image)

w.r.t figs shown \[ \frac{E_1}{I_2} = \frac{E_2}{I_1} = ZT \]

If \( E_1 = E_2 \) then \( I_1 = I_2 \).

**Thevinin’s and Norton’s Theorems:**

If we are interested in the solution of the current or voltage of a small part of the network, it is convenient from the computational point of view to simplify the network, except that small part in question, by a simple equivalent. This is achieved by Thevinin’s Theorem or Norton’s theorem.

**Thevinin’s Theorem:**

If two linear networks one M with passive elements and sources and the other N with passive elements only and there is no magnetic coupling between M and N, are connected together at terminals A and B, then with respect to terminals A and B, the network M can
be replaced by an equivalent network comprising a single voltage source in series with a single impedance. The single voltage source is the open circuit voltage across the terminals A and B and single series impedance is the impedance of the network M as viewed from A and B with independent voltage sources short circuited and independent current sources open circuited. Dependent sources if any are to be retained.

Arrange the networks M and N such that N is the part of the network where response is required.

To prove this theorem, consider the circuit shown in Fig.

Suppose the required response is the current $I_L$ in $Z_L$. Connected between A and B. According to Thevinin’s theorem the following steps are involved to calculate $I_L$

Step 1:

Remove $Z_L$ and measure the open circuit voltage across AB. This is also called as Thevinin’s voltage and is denoted as $V_{TH}$
Step 2:

To obtain the single impedance as viewed from A and B, replace the network in Fig. replacing the sources. This single impedance is called Thevinin’s Impedance and is denoted by $Z_{TH}$

\[
Z_{TH} = Z_{1} + Z_{2} + Z_{S}
\]

\[
V_{TH} = V_{AB} = \frac{E_{1} - I_{S}Z_{S}}{Z_{1} + Z_{2} + Z_{S}} \cdot Z_{1} + E_{2}
\]

\[
V_{TH} = V_{AB} = \frac{(E_{1} + E_{2})(Z_{1} + Z_{2} + Z_{S}) - (E_{1} - I_{S}Z_{S})Z_{1}}{Z_{1} + Z_{2} + Z_{S}}
\]

Step 3:

\[
Z_{TH} = \frac{Z_{1}(Z_{2} + Z_{S})}{Z_{1} + Z_{2} + Z_{S}}
\]
Write the thevinin’s network and re introduce ZL

Then the current in ZL is

\[ I_L = \frac{V_{TH}}{Z_{TH} + Z_L} \]

\[ = \frac{(E_1 + E_2)(Z_1 + Z_2 + Z_S) - (E_1 - I_SZ_S)Z_1}{Z_1(Z_2 + Z_S) + Z_L + Z_2(Z_1 + Z_2 + Z_S)} \]

To verify the correctness of this, write loop equations for the network to find the current in ZL

\[
\begin{vmatrix}
(E_1 + E_2) & Z_1 \\
(E_1 - I_SZ_S) & Z_1 + Z_2 + Z_S \\
Z_1 + Z_L & Z_1 \\
Z_1 & Z_1 + Z_2 + Z_S
\end{vmatrix}
\]
Norton’s Theorem:

The Thevinin’s equivalent consists of a voltage source and a series impedance. If the circuit is transformed to its equivalent current source, we get Norton’s equivalent. Thus Norton’s theorem is the dual of the Thevinin’s theorem.

If two linear networks, one M with passive elements and sources and the other N with passive elements only and with no magnetic coupling between M and N, are connected together at terminals A and B, then with respect to terminals A and B, the network M can be replaced by a single current source in parallel with a single impedance. The single current source is the short circuit current in AB and the single impedance is the impedance of the network M as viewed from A and B with independent sources being replaced by their internal impedances.

The proof of the Norton’s theorem is simple.

Consider the same network that is considered for the Thevinin’s Theorem and for the same response.

Step 1: Short the terminals A and B and measure the short circuit current in AB, this is Norton’s current source.

\[
I_N = I_{sc} = \frac{E_1 + E_2}{Z_1} + \frac{E_2 + I_s Z_S}{Z_2 + Z_S}
\]
\[ \frac{(E_1 + E_2)(Z_2 + Z_s) + (E_2 + I_s Z_s)Z_1}{Z_1(Z_2 + Z_s)} \]

Step 2: This is the same as in the case of thevinin’s theorem

Step 3: write the Norton’s equivalent and reintroduce \( Z_L \)

Then the current in \( Z_L \) is

\[
I_L = I_N \cdot \frac{Z_n}{Z_n + Z_L} - \frac{(E_1 + E_2)(Z_2 + Z_s) + (E_2 + I_s Z_s)Z_1}{Z_1(Z_2 + Z_s) + Z_L(Z_1 + Z_2 + Z_s)}
\]

Verification is to be done as in Thevinin’s Theorem

Determination of Thevinin’s or Norton’s equivalent when dependent sources are present

Since

\[
I_L = \frac{V_{TH}}{Z_{TH} + Z_L} = \frac{I_N \cdot Z_{TH}}{Z_{TH} + Z_L}
\]

\( Z_{TH} \) can also be determined as

\[
Z_{TH} = \frac{V_{TH}}{I_N} = \frac{\text{o.c voltage across AB}}{\text{s.c current in AB}}
\]

When network contains both dependent and independent sources. It is convenient to determine \( Z_{TH} \) by finding both the open circuit voltage and short circuit current

If the network contains only dependent sources both \( V_{TH} \) and \( I_N \) are zero in the absence of independent sources. Then apply a constant voltage source (or resultant source) and the ratio of voltage to current gives the \( Z_{TH} \). However there cannot be an independent source ie, \( V_{TH} \) or \( I_N \) in the equivalent network.
Maximum Transfer Theorem:-

When a linear network containing sources and passive elements is connected at terminals A and B to a passive linear network, maximum power is transferred to the passive network when its impedance becomes the complex conjugate of the Thevinin’s impedance of the source containing network as viewed form the terminals A and B.

Fig represents a network with sources replaced by its Thevinin’s equivalent of source of $E_{Th}$ volts and impedance $Z_s$, connected to a passive network of impedance $z$ at terminals A & B. With $Z_s = R_s + jX_s$ and $z = R + jX$, The proof of the theorem is as follows

Current in the circuit is

$$I = \frac{E_{Th}}{\sqrt{(R_s + R)^2 + (X_s + X)^2}}$$

The power delivered to the load is

$$P = I^2R = \frac{E_{Th}^2}{(R_s + R)^2 + (X_s + X)^2} \cdot R$$

As $P = \int (R, X)$ and since $P$ is maximum when $dP = 0$

We have

$$\frac{\delta P}{\delta R} \cdot dR + \frac{\delta P}{\delta X} \cdot dX = 0$$

power is maximum when $\frac{\delta P}{\delta R} = 0$ and $\frac{\delta P}{\delta X} = 0$ simultaneously

$$\frac{\delta P}{\delta R} = \frac{(R_s + R)^2 + (X_s + X)^2 - R \{2(R_s + R)\}}{D^2} = 0$$

$$(R_s + R)^2 + (X_s + X)^2 - 2R\{2(R_s + R)\} = 0 \quad (4)$$

$$\frac{\delta P}{\delta X} = -R\{2(X_s + X)\} = 0$$

$$2(R_s + X) = 0\quad (5)$$

From (5) we have $X = -X_s \quad (6)$

Substituting in (4) $(R_s + R)^2 = 2R(R_s + R)$, $R_s + R = 2R$

ie, $R = R_s$

Alternatively as $P = \frac{E^2R}{(R_s + R)^2 + (X_s + X)^2}$
\[(Rs+R)^2+(Xs+X)^2\]

\[= \frac{E^2Z\cos\Theta}{(Rs+Z\cos\Theta)^2+(Xs+Z\sin\Theta)^2}\]

\[= \frac{E^2Z\cos\Theta}{Zs^2+Z^2+2ZZs\cos(\Theta-\Theta_s)}\] (7)

ie \(P=f(Z,\Theta)\)

\[dP = \frac{\delta P}{\delta Z}\cdot dZ + \frac{\delta P}{\delta \Theta}\cdot d\Theta = 0\]

for \(P_{max}\)

\[\frac{\delta P}{\delta Z} = 0 = \{Z_s^2+Z^2+2Z\cdot Z_s\cos(\Theta-\Theta_s)\}\cdot \cos\Theta - Z\cdot \cos\Theta \cdot \{2Z+2Z_s\cos(\Theta-\Theta_s)\}\]

ie \(Z_s^2+Z^2+2Z\cdot Z_s\cos(\Theta-\Theta_s)\). Or \(|Z|=|Z_s|\)

\[\delta \Theta \]

then with

\[\frac{\delta P}{\delta \Theta} = 0 = \{Z_s^2+Z^2+2Z\cdot Z_s\cos(\Theta-\Theta_s)\}\cdot Z(-\sin\Theta)-Z\cdot \cos\Theta \cdot \{ZS^2+Z^2+2Z\cdot Z_s\sin(\Theta-\Theta_s)\}\]

\[\delta \Theta \]

\[(Z_s^2+Z^2)\sin\Theta = 2Z\cdot Z_s\cdot \{\cos\Theta \cdot \sin(\Theta-\Theta_s) - \sin\Theta \cdot \cos(\Theta-\Theta_s)\}\]

\[= -2Z\cdot Z_s\cdot \sin\Theta_s\] (9)

Substituting (8) in (9)

\[2Z\cdot \sin\Theta = -2Z_s^2\cdot \sin\Theta_s\]

\[\Theta = -\Theta_s\]

\[Z\cdot \Theta = Z_s\cdot -\Theta_s\]

Efficiency of Power Transfer:

With \(R_s=R_L\) and \(X_s=-X_L\) Substituting in (1)

\[P_{L_{max}} = \frac{E^2_{TH}R}{(2R)^2} = \frac{E^2_{TH}}{4R}\]

and the power supplied is \(P_s = \frac{E^2_{TH}}{(2R)^2} \cdot \frac{2R}{4R} = \frac{E^2_{TH}}{2R}\)

Then \(\eta_{tra} = \frac{P_L}{P_s} = \frac{E^2_{TH}}{E^2_{TH}/2R} = \frac{2}{2} = 1 \cdot 50\%\)

This means to transmit maximum power to the load 50% power generated is the loss. Such a low efficiency cannot be permitted in power systems involving large blocks of power where \(R_L\) is very
large compared to $R_s$. Therefore constant voltage power systems are not designed to operate on the basis of maximum power transfer.

However in communication systems the power to be handled is small as these systems are low current circuits. Thus impedance matching is considerable factor in communication networks.

However between $R$ & $X$ if either $R$ or $X$ is restricted and between $Z$ and $\Theta$ if either $|Z|$ or $\Theta$ is restricted the conditions for Max P is stated as follows

Case (i):- $R$ of $Z$ is varied keeping $X$ constant with $R$ only Variable, conditions for max power transfer is $(R_s+R)^2+(X_s+X)^2-2R(R_s+R)=0$  

$R_s^2+R^2+2R_sR+(X_s+X)^2-2R_sR-2R^2=0$  

$R=\sqrt{R_s^2+(X_s+X)^2}$

Case (ii):- If $Z$ contains only $R$ i.e., $x=0$ then from the eqn derived above 

$R=|Z_s|\sqrt{R_s^2+X_s^2}$

Case (iii):- If $|Z|$ is varied keeping $\Theta$ constant then from (8) $|Z|=|Z_s|$ 

Case (iv):- If $|Z|$ is constant but $\Theta$ is varied

Then from eqn (9) $(Z^2+Z_s^2)$ Sin $\Theta=-2Z Z_s$ Sin $\Theta$

$\text{Sin}\Theta = \frac{-2ZZ_s}{(Z^2+Z_s^2)}$

Then power transfer to load may be calculated by substituting for $R$ and $X$ for specified condition. For example

For case(ii) $P_{\text{max}}$ is given by 

$P_{\text{max}} = \frac{E^2R}{(R_s+R)^2+(X_s+X)^2}$  

$= \frac{E^2Z_s}{(R_s+Z_s)^2+X_s^2} = \frac{E^2Z_s}{R_s^2+2R_sZ_s+Z_s^2+X_s^2}$

$= \frac{E^2}{2(Z_s+R_s)}$ (ie $R_s^2+X_s^2=Z_s^2$)
Millman’s Theorem:

Certain simple combinations of potential and current source equivalents are of use because they offer simplification in solutions of more extensive networks in which combinations occur. Millman’s Theorem says that “if a number of voltage sources with internal impedances are connected in parallel across two terminals, then the entire combination can be replaced by a single voltage source in series with single impedance”.

The single voltage is the ratio

\[
\frac{\text{Sum of the product of individual voltage sources and their series admittances}}{\text{Sum of all series admittances}}
\]

and the single series impedance is the reciprocal of sum of all series admittances.

Let \( E_1, E_2, \ldots, E_n \) be the voltage sources and \( Z_1, Z_2, \ldots, Z_n \) are their respective impedances. All these are connected between A & B with \( Y=1/Z \), according to Millman’s Theorem, the single voltage source that replaces all these between A & B is

\[
E_{AB} = \frac{\sum_{K=1}^{n} E_K Y_K}{\sum_{K=1}^{n} Y_K}
\]

And

The single impedance is

\[
Z = \frac{1}{\sum_{K=1}^{n} Y_K}
\]
Proof: Transform each voltage into its equivalent current source. Then the circuit is as in Fig.

With \( Y = \frac{1}{Z} \) the circuit is simplified as
\[
E_1 Y_1 + E_2 Y_2 + \ldots + E_n Y_n = \sum E_K Y_K
\]

Which is a single current source in series with a single admittance.

Retransforming this into the equivalent voltage source

The theorem can be stated as “If a number of current sources with their parallel admittances are connected in series between terminals A and B, then they can be replaced by a single current source in parallel with a single admittance. The single current source is the ratio

\[
\frac{\sum E Y}{\sum Y} = \frac{1}{Z}
\]

And the single shunt admittance is the reciprocal of the sum of all shunt impedances.

Let \( I_1, I_2, \ldots, I_n \) be the n number of current sources and \( Y_1, Y_2, \ldots, Y_n \) be their respective shunt admittances connected in series between A & B. Then according to Millman’s
Theorem they can be replaced by single current $I_{AB}$ in parallel with a single admittance $Y_{AB}$ where

$$I_{AB} = \sum I_K Z_K$$

And

$$Y_{AB} = \frac{1}{\sum Z_K}$$

Transform each current source into its equivalent voltage source to get the circuit as in fig

$$\begin{align*}
I_1 & = \frac{I_1}{Y_1} \\
I_2 & = \frac{I_2}{Y_2} \\
\vdots & = \frac{I_n}{Y_n}
\end{align*}$$

Retransforming to equivalent current source

$$\begin{align*}
I_{AB} & = \frac{\sum I_K Z_K}{\sum Z_K} \\
& = \frac{1}{\sum Z_K}
\end{align*}$$
**TWO PORT PARAMETERS**

**PORT**:- Pair of terminals at which an electrical signal enters or leaves a network.

**One port network**:- Network having only one port.
Ex: Domestic appliances, Motor, Generator, Thevinin’s or Norton networks

**Two port network**:- Network having an input port and an output port.
Ex: Amplifiers, Transistors, communication circuits, Power transmission & distribution lines
Filters, attenuators, transformers etc

**Multi port network**:- Network having more than two ports.
Ex: Power Transmission lines, Distributions Lines, Communication lines.

Two port networks act as building blocks of electrical or electronic circuits such as electronic systems, communication circuits, control systems and transmission & distribution systems. A one port or two port network can be connected with another two port network either in cascade, series or in parallel. In Thevinins or Nortons networks, we are not interested in the detailed working of a major part of the network. Similarly it is not necessary to know the inner working of the two port network but by measuring the voltages and currents at input and at output port, the network can be characterized with a set of parameters to predict how a two port network interact with other networks.
networks. Often the circuit between the two ports is highly complex. The two port parameters provide a shorthand method for analyzing the input-output properties of two ports without having to deal directly with the highly complex circuit internal to the two port.

These networks are linear and passive and may contain controlled sources but not independent sources.

While defining two port parameters we put the condition that one of the ports is either open circuited or short circuited.

In these networks there are four variables \(V_1, I_1\) and \(V_2, I_2\). Two of them are expressed in terms of the other two, to define two port parameters.

<table>
<thead>
<tr>
<th>Sl. No.</th>
<th>Parameters</th>
<th>Dependent Variable</th>
<th>Independent Variable</th>
<th>Equations</th>
</tr>
</thead>
</table>
| 1.     | \(z\) Parameters | \(V_1, V_2\) | \(I_1, I_2\) | \[
\begin{bmatrix}
V_1 \\
V_2
\end{bmatrix} =
\begin{bmatrix}
z_{11} & z_{12} \\
z_{21} & z_{22}
\end{bmatrix}
\begin{bmatrix}
I_1 \\
I_2
\end{bmatrix}
\] |
| 2.     | \(y\) parameters | \(I_1, I_2\) | \(V_1, V_2\) | \[
\begin{bmatrix}
I_1 \\
I_2
\end{bmatrix} =
\begin{bmatrix}
y_{11} & y_{12} \\
y_{21} & y_{22}
\end{bmatrix}
\begin{bmatrix}
V_1 \\
V_2
\end{bmatrix}
\] |
| 3.     | \(h\) parameters | \(V_1, I_2\) | \(I_1, V_2\) | \[
\begin{bmatrix}
V_1 \\
I_2
\end{bmatrix} =
\begin{bmatrix}
h_{11} & h_{12} \\
h_{21} & h_{22}
\end{bmatrix}
\begin{bmatrix}
I_1 \\
V_2
\end{bmatrix}
\] |
| 4.     | \(t\) parameters | \(V_1, I_1\) | \(V_2, I_2\) | \[
\begin{bmatrix}
V_1 \\
I_1
\end{bmatrix} =
\begin{bmatrix}
A & B \\
C & D
\end{bmatrix}
\begin{bmatrix}
V_2 \\
I_2
\end{bmatrix}
\] |

**DEFINITIONS**

(1) \(Z\) parameters (open circuit impedance parameters)

\[V_1 = z_{11}I_1 + z_{12}I_2\]

\[
z_{11} = \frac{V_1}{I_1} \quad I_2 = 0
\]

\[
z_{12} = \frac{V_1}{I_2} \quad I_1 = 0
\]
\[ V_2 = z_{21} I_1 + z_{22} I_2 \]

\[ z_{21} = \frac{V_2}{I_1} \quad I_2 = 0 \]

\[ z_{22} = \frac{V_2}{I_2} \quad I_1 = 0 \]

For \( z_{11} \) and \( z_{21} \) - output port opened
\( z_{12} \) and \( z_{22} \) - input port opened

Hence the name open circuit impedance parameters

Equivalent networks in terms of controlled sources:

Network (i)

Network (ii) By writing

\[ V_1 = (z_{11} - z_{12}) I_1 + z_{12} (I_1 + I_2) \]

\[ V_2 = (z_{21} - z_{12}) I_1 + (z_{22} - z_{12}) I_2 + z_{12} (I_1 + I_2) \]

The \( z \) parameters simplify the problem of obtaining the characteristics of two 2 port networks connected in series

**(2) \( y \) parameters**

\[ I_1 = y_{11} V_1 + y_{12} V_2 \]

\[ y_{11} = \frac{I_1}{V_1} \]

\[ y_{12} = \frac{I_1}{V_2} \]

\[ y_{21} = \frac{I_2}{V_1} \]

\[ y_{22} = \frac{I_2}{V_2} \]

For \( y_{11} \) and \( y_{21} \) - port 2 is shorted
\( y_{12} \) and \( y_{22} \) - port 1 is shorted

Hence they are called short circuit admittance parameters
Equivalent networks in terms of controlled sources

(ii) by writing

\[ I_1 = (y_{11} + y_{12}) V_1 - y_{12} (V_1 + V_2) \]
\[ I_2 = (y_{21} - y_{12}) V_1 + (y_{22} + y_{12}) V_2 - y_{12} (V_2 - V_1) \]

The \( y \) parameters are very useful to know the characteristics of two 2 port Networks connected in parallel

**Hybrid parameters:**

\[ V_1 = h_{11} I_1 + h_{12} V_2 \]
\[ h_{11} = \left| \frac{V_1}{I_1} \right| V_2 = 0 \]
\[ h_{12} = \left| \frac{V_1}{V_2} \right| I_1 = 0 \]

\[ I_2 = h_{21} I_1 + h_{22} V_2 \]
\[ h_{22} = \left| \frac{I_2}{I_1} \right| V_2 = 0 \]
\[ h_{22} = \left| \frac{I_2}{V_2} \right| I_1 = 0 \]
Parameter values for bipolar junction transistors are commonly quoted in terms of h parameters.

**Transmission or ABCD parameters**

\[ V_1 = A V_2 - B I_2 \]
\[ I_1 = C V_2 - D I_2 \]
\[ A = \frac{V_1}{V_2} \quad I_2 = 0 \]
\[ B = \frac{V_1}{-I_2} \quad V_2 = 0 \]
\[ C = \frac{I_1}{V_2} \quad I_2 = 0 \]
\[ D = \frac{I_1}{-I_2} \quad V_2 = 0 \]

As the name indicates the major use of these parameters arise in transmission line analysis and when two 2 ports are connected in cascade.

**Relationship between two port parameters**:

Relationship between different two port parameters can be obtained as follows. From the given set of two port parameters, rearrange the equations collecting terms of dependent variables of new set of parameters to the left. Then form matrix equations and from matrix manipulations obtain the new set in terms of the given set.

(i) Relationship between \( z \) and \( y \) parameters for \( x \) parameters

\[ \begin{bmatrix} V \end{bmatrix} = \begin{bmatrix} z \end{bmatrix} \begin{bmatrix} I \end{bmatrix} \]

Then

\[ \begin{bmatrix} I_1 \\ I_2 \end{bmatrix} = \begin{bmatrix} z_{11} & z_{12} \\ z_{21} & z_{22} \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} \]

\[ = \frac{1}{\Delta_z} \begin{bmatrix} z_{22} & -z_{12} \\ -z_{21} & z_{11} \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} \]

where \( \Delta_z = z_{11} z_{22} - z_{12} z_{21} \)

\[ \begin{bmatrix} y_{11} & y_{12} \\ y_{21} & y_{22} \end{bmatrix} = \frac{1}{\Delta_z} \begin{bmatrix} \frac{z_{22}}{\Delta_z} & -\frac{z_{12}}{\Delta_z} \\ -\frac{z_{21}}{\Delta_z} & \frac{z_{11}}{\Delta_z} \end{bmatrix} \]

Similarly

\[ \begin{bmatrix} z_{11} & z_{12} \\ z_{21} & z_{22} \end{bmatrix} = \frac{1}{\Delta_y} \begin{bmatrix} y_{22} - y_{12} \\ -y_{21} & y_{22} \end{bmatrix} \]

(ii) Relationship between \( y \) and \( h \)

From

\[ I_1 = y_{11} V_1 + y_{12} V_2 \]
\[ I_2 = y_{21} V_1 + y_{22} V_2 \]

Rearranging

\[ y_{11} V_1 = I_1 - y_{12} V_2 \]
\[ y_{21} V_1 - I_2 = -y_{22} V_2 \]

\[ \begin{bmatrix} y_{11} & 0 \\ y_{21} & -1 \end{bmatrix} \begin{bmatrix} V_1 \end{bmatrix} = \begin{bmatrix} 1 & -y_{12} \\ 0 & -y_{22} \end{bmatrix} \begin{bmatrix} I_1 \\ V_2 \end{bmatrix} \]
(iii) To Express T-parameters in terms of h-Parameters:

Equations for T-parameters,

\[ V_1 = AV_2 - BI_2 \]
\[ I_1 = CV_2 - DI_2 \]

\[ \begin{align*}
    V_1 &= [y_{11} \quad y_{12}]^{-1} \begin{bmatrix} 1 & -y_{12} \end{bmatrix} I_1 \\
    I_2 &= [y_{21} \quad -1] \begin{bmatrix} 0 & -y_{22} \end{bmatrix} V_2 \\
    &= -1 \begin{bmatrix} -1 & 0 \end{bmatrix} \begin{bmatrix} 1 & -y_{12} \end{bmatrix} I_1 \\
    &= -1 \begin{bmatrix} -1 & y_{12} \end{bmatrix} \begin{bmatrix} y_{11} \quad -y_{21} \end{bmatrix} \begin{bmatrix} 1 \quad 0 \end{bmatrix} \begin{bmatrix} 1 \quad -y_{12} \end{bmatrix} I_1 \\
    &= -1 \begin{bmatrix} 1 & -y_{12} \\
    y_{11} & y_{12} \end{bmatrix} \begin{bmatrix} 1 \quad 0 \end{bmatrix} \begin{bmatrix} I_1 \\
    y_{21} & y_{22} \end{bmatrix} V_2 \end{align*} \]

\[ \begin{bmatrix} h_{11} & h_{12} \\
    h_{21} & h_{22} \end{bmatrix} = \begin{bmatrix} 1 & -y_{12} \\
    y_{11} & y_{12} \end{bmatrix} \begin{bmatrix} y_{11} \quad -y_{21} \end{bmatrix} \begin{bmatrix} 1 \quad 0 \end{bmatrix} \begin{bmatrix} I_1 \\
    y_{21} & y_{22} \end{bmatrix} V_2 \]

By a similar procedure, the relationship between any two sets of parameters can be established. The following table gives such relationships:

<table>
<thead>
<tr>
<th>Y</th>
<th>z</th>
<th>H</th>
<th>T</th>
</tr>
</thead>
</table>
| \[y\] | \[y_{11} \quad y_{12} \]
| \[y_{21} \quad y_{22} \] | \[z_{22} \quad -z_{12} \]
| \[\Delta_z \quad \Delta_z \] | \[1 \quad -h_{12} \]
| \[h_{11} \quad h_{11} \] | \[-h_{12} \quad -h_{12} \]
| \[\frac{D}{B} \quad \Delta_z \] | \[-\frac{A}{B} \quad \Delta_z \]
| \[\frac{D}{B} \quad \Delta_z \] | \[-\frac{A}{B} \quad \Delta_z \]

<table>
<thead>
<tr>
<th>z</th>
</tr>
</thead>
</table>
| \[z_{11} \quad z_{12} \]
| \[\Delta_z \quad \Delta_z \] | \[\frac{1}{h_{11}} \quad \frac{h_{12}}{h_{12}} \]
| \[\frac{h_{11}}{h_{11}} \quad \frac{h_{12}}{h_{12}} \] | \[\frac{A}{C} \quad \Delta_z \]
| \[\frac{h_{11}}{h_{11}} \quad \frac{h_{12}}{h_{12}} \] | \[\frac{A}{C} \quad \Delta_z \]

By a similar procedure, the relationship between any two sets of parameters can be established. The following table gives such relationships:
Computations of Two Port Parameters:

A. By direct method i.e. using definitions

For z parameters, open output port (I2=0) find V1 & V2 in terms of I1 by equations

\[ Z_{11} = \frac{V_1}{I_1} \quad \text{and} \quad Z_{21} = \frac{V_2}{I_1} \]

Open input port (I1=0) find V1 & V2 in terms of I2. Calculate \( Z_{12} = \frac{V_1}{I_2} \) and \( Z_{22} = \frac{V_2}{I_2} \).

Similar procedure may be followed for y parameters by short circuiting the ports h & t parameters may be obtained by a combination of the above procedures.

B. \( z \) and \( y \) parameters: By node & mesh equations in standard form

For a reciprocal network (passive without controlled sources) with only two current Sources at input and output nodes, the node equations are

\[
\begin{align*}
I_1 &= Y_{11} V_1 + Y_{12} V_2 + Y_{13} V_3 + \cdots + Y_{1n} V_n \\
I_2 &= Y_{21} V_1 + Y_{22} V_2 + Y_{23} V_3 + \cdots + Y_{2n} V_n \\
0 &= Y_{31} V_1 + Y_{32} V_2 + Y_{33} V_3 + \cdots + Y_{3n} V_n \\
0 &= Y_{n1} V_1 + Y_{n2} V_2 + Y_{n3} V_3 + \cdots + Y_{nn} V_n
\end{align*}
\]

then \( V_1 = \frac{\Delta_{11}}{\Delta} I_1 + \frac{\Delta_{21}}{\Delta} I_2 \), where \( \Delta \) is the determinant of the \( Y \) matrix.

\[
V_2 = \frac{\Delta_{12}}{\Delta} I_1 + \frac{\Delta_{22}}{\Delta} I_2 \quad \Delta_{ij} \text{ cofactor of } Y_{ij} \text{ of } \Delta
\]

Comparing these with the \( z \) parameter equations,

we have \( z_{11} = \frac{\Delta_{11}}{\Delta} \) \( z_{22} = \frac{\Delta_{22}}{\Delta} \) \( z_{12} = \frac{\Delta_{21}}{\Delta} \) \( z_{21} = \frac{\Delta_{12}}{\Delta} \)

Similarly for such networks, the loop equations with voltage sources only at port 1 and 2

\[
\begin{align*}
V_1 &= Z_{11} I_1 + Z_{12} I_2 + \cdots + Z_{1m} I_m \\
V_2 &= Z_{21} I_1 + Z_{22} I_2 + \cdots + Z_{2m} I_m \\
0 &= \quad \cdots \quad \cdots \quad \cdots \quad \cdots \quad \cdots \\
O &= Z_{m1} I_1 + Z_{m2} I_2 + \cdots + Z_{mm} I_m
\end{align*}
\]
Then

\[ I_1 = \frac{D_{11}}{D} V_1 + \frac{D_{21}}{D} V_2 \]
\[ I_2 = \frac{D_{12}}{D} V_1 + \frac{D_{22}}{D} V_2 \]

where D is the determinant of the Z matrix and D_{ij} is the co-factor of the element Z_{ij} of Z matrix.

Comparing these with \([y]\) equations

Thus we have

\[ y_{11} = \frac{D_{11}}{D} \]
\[ y_{22} = \frac{D_{22}}{D} \]
\[ y_{12} = \frac{D_{12}}{D} \]
\[ y_{22} = \frac{D_{22}}{D} \]

Alternative methods

For \(z\) parameters the mesh equations are

\[ V_1 = Z_{11} I_1 + Z_{12} I_2 + \ldots + Z_{1m} I_m \]
\[ V_2 = Z_{21} I_1 + Z_{22} I_2 + \ldots + Z_{2m} I_m \]
\[ O = \ldots \ldots \]
\[ O = Z_{m1} I_1 + Z_{m2} I_2 + \ldots + Z_{mn} I_m \]

By matrix partitioning the above equations can be written as

\[
\begin{bmatrix}
V_1 \\
V_2 \\
0
\end{bmatrix} =
\begin{bmatrix}
Z_{11} & Z_{12} & \ldots & Z_{1m} \\
Z_{21} & Z_{22} & \ldots & Z_{2m} \\
\vdots & \vdots & \ddots & \vdots \\
Z_{n1} & Z_{n2} & \ldots & Z_{nm}
\end{bmatrix}
\begin{bmatrix}
I_1 \\
I_2 \\
\vdots \\
I_n
\end{bmatrix}
\]

\[
\begin{bmatrix}
V_1 \\
V_2 \\
0
\end{bmatrix} =
\begin{bmatrix}
M & N
\end{bmatrix}
\begin{bmatrix}
I_1 \\
I_2 \\
\vdots \\
I_n
\end{bmatrix}
\]

\[
\begin{bmatrix}
V_1 \\
V_2
\end{bmatrix} =
\begin{bmatrix}
M - NQ^{-1}P
\end{bmatrix}
\begin{bmatrix}
I_1 \\
I_2
\end{bmatrix}
\]
Similarly for Y parameters

\[
\begin{bmatrix}
I_1 \\
I_2 \\
0
\end{bmatrix} =
\begin{bmatrix}
Y_{11} & Y_{12} & \cdots & Y_{1n} \\
Y_{21} & Y_{22} & \cdots & Y_{2n} \\
\vdots & \vdots & \ddots & \vdots \\
Y_{n1} & Y_{n2} & \cdots & Y_{nn}
\end{bmatrix}
\begin{bmatrix}
V_1 \\
V_2 \\
\vdots \\
V_n
\end{bmatrix}
\]

\[
\begin{bmatrix}
I_1 \\
I_2 \\
0
\end{bmatrix} =
\begin{bmatrix}
M & N \\
- & - \\
O & P & Q
\end{bmatrix}
\begin{bmatrix}
V_1 \\
V_2 \\
V_3
\end{bmatrix}
\]

\[
\begin{bmatrix}
I_1 \\
I_2
\end{bmatrix} = 
\begin{bmatrix}
M - NQ^{-1}P
\end{bmatrix}
\]

C. By reducing the network (containing passive elements only) to single T or D by T-D transformations

If the network is reduced to a T network as shown

\[
\begin{align*}
&\begin{array}{ccc}
+ & Z_1 & + \\
\hline
V_1 & I_1 & Z_2 \\
\hline
- & Z_3 & -
\end{array} \\
&I_2
\end{align*}
\]

Then

\[
\begin{align*}
V_1 &= (Z_1 + Z_3)I_1 + Z_3I_2 \\
V_2 &= Z_1I_1 + (Z_2 + Z_3)I_2
\end{align*}
\]

from which

\[
\begin{align*}
z_{11} &= Z_1 + Z_3 \\
z_{22} &= Z_2 + Z_3 \\
z_{12} &= z_{21} = z_{13}
\end{align*}
\]

If the network is brought to Π network as shown

\[
\begin{align*}
&\begin{array}{ccc}
+ & Y_3 & + \\
\hline
V_1 & Y_1 & Y_2 \\
\hline
- & & -
\end{array} \\
&I_1 & I_2
\end{align*}
\]

Then

\[
\begin{align*}
I_1 &= (Y_1 + Y_3)V_1 - Y_3V_2 \\
I_2 &= -Y_3V_1 + (Y_2 + Y_3)V_2
\end{align*}
\]

from which

\[
\begin{align*}
y_{11} &= Y_1 + Y_3 \\
y_{22} &= Y_2 + Y_3 \\
y_{12} &= y_{21} = -Y_3
\end{align*}
\]
t-PARAMETERS FOR T & π NETWORKS

For a T network the mesh equations are:

\[
\begin{align*}
V_1 &= (Z_1 + Z_2)I_1 + I_2Z_3 \\
V_2 &= Z_1I_1 + (Z_2 + Z_3)I_2 \\
\end{align*}
\]

Re arranging

\[
\begin{align*}
V_1 - (Z_1 + Z_3)I_1 &= I_2Z_3 \\
-Z_1I_1 &= (Z_2 + Z_3)I_2 - V_2 \\
\end{align*}
\]

In matrix form

\[
\begin{bmatrix}
V_1 \\
I_1 \\
\end{bmatrix} = \begin{bmatrix}
1 & -(Z_1 + Z_3) \\
0 & -Z_3 \\
\end{bmatrix}^{-1} \begin{bmatrix}
0 & Z_3 \\
-1 & Z_2 + Z_3 \\
\end{bmatrix} \begin{bmatrix}
V_2 \\
-I_2 \\
\end{bmatrix}
\]

\[
= -\frac{1}{Z_3} \begin{bmatrix}
-Z_3 & Z_1 + Z_3 \\
0 & 1 \\
\end{bmatrix} \begin{bmatrix}
0 & Z_3 \\
-1 & Z_2 + Z_3 \\
\end{bmatrix} \begin{bmatrix}
V_2 \\
-I_2 \\
\end{bmatrix}
\]

\[
= \frac{1}{Z_3} \begin{bmatrix}
(Z_1 + Z_3) & Z_1Z_2 + Z_2Z_3 + Z_3Z_1 \\
1 & Z_2 + Z_3 \\
\end{bmatrix} \begin{bmatrix}
V_2 \\
-I_2 \\
\end{bmatrix}
\]

\[
\begin{bmatrix}
A & B \\
C & D \\
\end{bmatrix} = \frac{1}{Z_3} \begin{bmatrix}
(Z_1 + Z_3) & Z_1Z_2 + Z_2Z_3 + Z_3Z_1 \\
1 & Z_2 + Z_3 \\
\end{bmatrix}
\]

For the π network shown, the equations are:

\[
\begin{align*}
I_1 &= (Y_1 + Y_3) V_1 - Y_3 V_2 \\
I_2 &= -V_1 Y_3 + (Y_2 + Y_3) V_2 \\
\end{align*}
\]

Re arranging

\[
(V_1 + Y_3)V_1 - I_1 = V_2 Y_3 \\
Y_3 V_1 = (Y_2 + Y_3) V_2 - I_2 \\
\]

In matrix form

\[
\begin{bmatrix}
V_1 \\
I_1 \\
\end{bmatrix} = \begin{bmatrix}
Y_1 + Y_3 & 1 \\
Y_3 & 0 \\
\end{bmatrix}^{-1} \begin{bmatrix}
0 & -Y_3 \\
Y_2 + Y_3 & -1 \\
\end{bmatrix} \begin{bmatrix}
V_2 \\
-I_2 \\
\end{bmatrix}
\]

\[
= \frac{1}{Y_3} \begin{bmatrix}
0 & -1 \\
-Y_3 & Y_1 + Y_3 \\
\end{bmatrix} \begin{bmatrix}
0 & -Y_3 \\
Y_2 + Y_3 & -1 \\
\end{bmatrix} \begin{bmatrix}
V_2 \\
-I_2 \\
\end{bmatrix}
\]

\[
= \frac{1}{Y_3} \begin{bmatrix}
(Y_2 + Y_3) & 1 \\
Y_1 Y_2 + Y_2 Y_3 + Y_3 Y_1 & (Y_1 + Y_3) \\
\end{bmatrix} \begin{bmatrix}
V_2 \\
-I_2 \\
\end{bmatrix}
\]
\[
\begin{bmatrix}
A & B \\
C & D
\end{bmatrix} = \frac{1}{Y_3} \begin{bmatrix}
Y_2 + Y_3 \\
Y_1Y_2 + Y_2Y_3 + Y_3Y_1 \\
Y_1 + Y_3
\end{bmatrix}
\]

D. The above methods are based on the assumptions that the network does not contain controlled sources. However irrespective of the presence of the controlled sources, network equations may be written and then by any elimination process variables other than \(V_1, V_2, I_1 & I_2\) are eliminated. Then resulting two equations are brought to the Required form of two port parameters by manipulation.

**SYMMETRICAL CONDITIONS**

A two port is said to be symmetrical if the ports can be interchanged without changing the port voltage and currents.

\[
i.e. \quad \left| \begin{array}{c}
V_1 \\
I_1
\end{array} \right| \left| \begin{array}{c}
I_2 \\
V_2
\end{array} \right| = 0 = \left| \begin{array}{c}
V_2 \\
I_2
\end{array} \right| \left| \begin{array}{c}
I_1 \\
V_1
\end{array} \right|
\Rightarrow z_{11} = z_{22}
\]

By using the relationship between \(z\) and other parameters we can obtain the conditions for Symmetry in terms of other parameters.

As \(z_{11}=z_{22}\), in terms of \(y\) we have \(y_{11}=z_{12}/dz & y_{22}=z_{1}/dz\). \(\therefore y_{11}=y_{22}\).

In terms of \(h\) parameters as \(z_{11}=\Delta h/h_{22} & z_{22}=1/h_{22}\) we have \(\Delta h=h_{11}h_{22}-h_{12}h_{21} = 1\).

In terms of \(t\) parameters as \(z_1=A/C & z_{22}=D/C\) the condition is \(A=D\)

**Reciprocity condition in terms of two port parameters**

For the two networks shown for

<table>
<thead>
<tr>
<th>Network</th>
<th>Condition</th>
</tr>
</thead>
<tbody>
<tr>
<td>Fig 1</td>
<td>(V_1 = V) (I_2 = -I_a) (V_2 = 0)</td>
</tr>
<tr>
<td>Fig 2</td>
<td>(V_2 = V) (I_1 = -I_b) (V_1 = 0)</td>
</tr>
</tbody>
</table>

Condition for reciprocity is \(I_a = I_b\)

From \(z\) parameters
\[
V_1 = z_{11}I_1 + z_{12}I_2 \\
V_2 = z_{21}I_1 + z_{22}I_2
\]

we have from fig(1)
\[
V = z_{11}I_1 - z_{12}I_a \\
O = z_{21}I_1 - z_{22}I_a
\]
\[:. I_a = \frac{-z_{21}V}{-\Delta z} = \frac{z_{12}V}{\Delta z} \]

From fig(2)
\[
O = z_{11}I_b + z_{12}I_2 \\
V = -z_{21}I_b + z_{12}I_2
\]
\[I_b = \frac{-z_{21}V}{-\Delta z} \]
then for \( I_a = I_b \)
\[z_{12} = z_{21} \]

For reciprocity with \( z_{12} = z_{21} \),
In terms of \( y \) parameters \( z_{12} = \frac{-y_{12}}{\Delta y} \) & \( z_{21} = \frac{-y_{21}}{\Delta y} \) condition is \( y_{12} = y_{21} \)
In terms of \( h \) parameters \( z_{12} = \frac{h_{12}}{h_{22}} \) & \( z_{21} = \frac{-h_{21}}{h_{22}} \) the condition is \( h_{12} = -h_{21} \)
In terms of \( t \) parameters \( z_{12} = \frac{\Delta t}{C} \) & \( z_{21} = \frac{1}{C} \) the condition is \( \Delta t = AD - BC = 1 \)

<table>
<thead>
<tr>
<th>Parameters</th>
<th>Condition for</th>
</tr>
</thead>
<tbody>
<tr>
<td></td>
<td>Reciprocity</td>
</tr>
<tr>
<td>( Z )</td>
<td>( Z_{12} = Z_{22} )</td>
</tr>
<tr>
<td>( y )</td>
<td>( y_{12} = y_{22} )</td>
</tr>
<tr>
<td>( h )</td>
<td>( h_{12} = -h_{21} )</td>
</tr>
<tr>
<td></td>
<td></td>
</tr>
<tr>
<td>( t )</td>
<td>( AD - BC = 1 )</td>
</tr>
</tbody>
</table>

CASCADE CONNECTION:-

In the network shown 2 two port networks are connected in cascade
For $N_a$, $[t] = \begin{bmatrix} A_a & B_a \\ C_a & D_a \end{bmatrix}$ for $N_b$, $[t] = \begin{bmatrix} A_b & B_b \\ C_b & D_b \end{bmatrix}$

For the resultant network $N$ $[t] = \begin{bmatrix} A & B \\ C & D \end{bmatrix}$

From the cascaded network we have

$$V_{la} = A_a V_{2a} - B_a I_{2a} \quad \text{for network } N_a$$

$$I_{la} = C_a V_{2a} - D_a I_{2a}$$

$$V_{lb} = A_b V_{2b} - B_b I_{2b} \quad \text{for network } N_b$$

$$I_{lb} = C_b V_{2b} - D_b I_{2b}$$

$$V_1 = AV_2 - BI_2 \quad \text{for network } N$$

$$I_1 = CV_2 - DI_2$$

From the network

$$I_1 = I_{la} + I_{2a} = -I_{lb} \quad I_2 = I_{2b}$$

$$V_1 = V_{la} \quad V_{2a} = V_{lb} \quad V_2 = V_{2b}$$

$$\begin{bmatrix} V_{la} \\ I_{la} \end{bmatrix} = \begin{bmatrix} A_a & B_a \\ C_a & D_a \end{bmatrix} \begin{bmatrix} -V_{2a} \\ -I_{2a} \end{bmatrix} \quad \text{and} \quad \begin{bmatrix} V_{lb} \\ I_{lb} \end{bmatrix} = \begin{bmatrix} A_b & B_b \\ C_b & D_b \end{bmatrix} \begin{bmatrix} -V_{2b} \\ -I_{2b} \end{bmatrix}$$

or

$$\begin{bmatrix} V_{la} \\ I_{la} \end{bmatrix} = \begin{bmatrix} A_a & B_a \\ C_a & D_a \end{bmatrix} \begin{bmatrix} V_{lb} \\ I_{lb} \end{bmatrix} = \begin{bmatrix} A_a & B_a \\ C_a & D_a \end{bmatrix} \begin{bmatrix} A_b & B_b \\ C_b & D_b \end{bmatrix} \begin{bmatrix} V_2 \\ -I_2 \end{bmatrix}$$

$$\begin{bmatrix} A & B \\ C & D \end{bmatrix} = \begin{bmatrix} A_a & B_a \\ C_a & D_a \end{bmatrix} \begin{bmatrix} A_b & B_b \\ C_b & D_b \end{bmatrix}$$

$$[T] = [T_a \mid T_b]$$
Writing the network from equations:

1) To write the network from mesh equations:
   
   With the equations in matrix form \([E]=[Z][I]\) and all meshes in clockwise direction, draw the graph of the network keeping in mind that there is a branch between mesh J and mesh K if \(Z_{jk}\) exists and the number of meshes is equal to the number of I’s. If \(Z_{jk}\) is zero, there is no branch common to meshes J and K.

For example, if the network contains 3 meshes with mutual \(Z\) exists among all the three, the graph is of the form as shown in fig 1.

On the other hand if there is no mutual \(Z\) between first and third meshes the graph is of the form as shown in fig 2

![Fig.1](image1)

![Fig.2](image2)

With this information

   Insert in each mutual line, the respective mutual \(Z\). (negative of \(Z_{jk}\))

   Insert in non-mutual line the sum \(Z_{KK} + Z_{Kj}\) for the \(K\)th mesh.

   Insert \(E_K\) in the non mutual line of mesh K. This is not unique since \(E_K\) can be split into many E’s and may be placed in many branches of the \(K\)th mesh or loop.

   Thus the network is obtained.

Problem

For the equations shown draw the network

\[
\begin{bmatrix}
35 - J1.62 & -10 + J31.83 & -20 - J15.71 \\
-10 + J31.83 & 35 - J25.55 & -20 \\
-20 - J15.71 & -20 & 50 + J15.71
\end{bmatrix}
\begin{bmatrix}
I_1 \\
I_2 \\
I_3
\end{bmatrix}
= \begin{bmatrix}
200 \\
0 \\
0
\end{bmatrix}
\]

Step-1

Step-2

Step-3
2) To write the network from node equations:

The number of equations indicate the number of independent nodes and non diagonal terms indicate –ve of the mutual admittances between the respective nodes. Zero in the non diagonal terms indicate no branch between the respective nodes. With this information draw the graph of the network including reference node.

Insert in each mutual branch the respective mutual admittance (-ve of the non diagonal term)

Insert in non mutual line the sum \( Y_{KK} + K_{Kj} \) for the \( K^{th} \) node.

Insert \( I'_K \) in the non mutual line of the node \( K \). This is not unique since \( I_K \) can be split into may I’s distributed in some of the other branches connected to node \( K \).

Thus obtain the network.

For example, if the network contains 3 independent nodes with mutual \( Y \) among all the three, the graph is of the form shown in fig 1

On the other hand if mutual \( Y \) exists between two nodes only then the graph is of the form shown in fig 2

![Fig. 1](image)

![Fig. 2](image)

In case the network contains, mixed sources and controlled sources, super meshes and super nodes are carefully identified.

**Problem**

For the equation shown draw the network

\[
\begin{bmatrix}
1 & -J1 & J1 \\
-J1 & J2 & -J1 \\
J1 & -J1 & 1
\end{bmatrix}
\begin{bmatrix}
V_1 \\
V_2 \\
V_3
\end{bmatrix}
= \begin{bmatrix}
1 \\
-1 \\
J1
\end{bmatrix}
\]

![Step - 1](image)

![Step - 2](image)

![Step - 3](image)